

Study on Partial Coating Approach for Reducing Cavity Backscattering

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Abstract – Cavities are strong scattering parts for aircrafts, while coating the interior surface of cavities is a general approach for reducing cavity backscattering. In this paper, the discontinuous Galerkin (DG) method based on the self-dual integral equation (SDIE) is employed to efficiently simulate scattering from partially coated cavities. The distance sparse preconditioner (DSP) is applied to DG-SDIE for speeding up convergence instead of the conventional block-diagonal preconditioner (BDP). An efficient partial coating scheme is presented. The numerical results of straight cavity, S-shaped cavity and complex aircraft cavity demonstrate our coating scheme can achieve similar effect of reducing backscattering by using smaller coated interior surface regions. Useful conclusions about partially coating the cavity are summarized.

Index Terms – Coated cavity, discontinuous Galerkin (DG) method, distance sparse preconditioner (DSP), self-dual integral equation (SDIE).

I. INTRODUCTION

Scattering from cavity structures often contributes significantly to the overall electromagnetic (EM) scattering from aircraft. A major way to reduce EM scattering from these cavities is to coat their interior surface with EM absorbing materials. How to efficiently coat the cavity becomes an attractive problem.

Method of moments (MoM) is a powerful tool for the EM scattering problem [1–3]. The thin coating surface is usually approximated by the impedance boundary condition (IBC) [4]. Formulations of MoM based on the surface integral equation and IBC were developed in [5–8]. However, these early formulations suffered from drawbacks of an ill-conditioned matrix. A robust and efficient self-dual integral equation (SDIE) was developed for scattering from coated objects [9]. A more efficient formulation was recently developed by employing the combined field integral equation (CFIE) and IBC, shortened to C-SDIE [10]. The discontinuous

Galerkin (DG) method was recently applied to C-SDIE (DG-C-SDIE) to further improve efficiency and flexibility for simulating scattering from non-uniform or partially coated cavities [11–14].

In this paper, DG-C-SDIE is employed to study scattering from partially coated cavities. A more efficient distance sparse preconditioner (DSP) is applied to DG-C-SDIE to replace the conventional block-diagonal preconditioner (BDP) used in [10]. Furthermore, a partial coating scheme is presented for more efficient coating. The numerical results of scattering from coated rectangular cavity, S-shaped curved cavity and realistic aircraft cavity demonstrate that our coating scheme can achieve the similar effects of reducing backscattering with a smaller coated surface region. Some useful results for partially coating a cavity to reduce backscattering are obtained and shown in section IV.

II. FORMULATION

Consider a coated cavity immersed in free space with permittivity ϵ_0 and permeability μ_0 . The cavity is illuminated by an incident plane wave ($\mathbf{E}^{inc}, \mathbf{H}^{inc}$). Electric current $\mathbf{J}$ and magnetic current $\mathbf{M}$ on the cavity surface S can be formulated by the CFIE as:

$$C(z_0 \mathbf{J}) - \frac{1}{2} z_0 \mathbf{J} + \hat{\mathbf{n}} \times C(\mathbf{M}) - \frac{1}{2} \hat{\mathbf{n}} \times \mathbf{M} = -\mathbf{e}^{inc} - \mathbf{j}^{inc}, \quad (1)$$

where $\hat{\mathbf{n}}$ denotes the outward normal vector of the cavity surface S , $z_0 = \sqrt{\epsilon_0/\mu_0}$, $\mathbf{e}^{inc} = \hat{\mathbf{n}} \times \mathbf{E}^{inc} \times \hat{\mathbf{n}}$, and $\mathbf{j}^{inc} = z_0 \hat{\mathbf{n}} \times \mathbf{H}^{inc}$. Operator $C = L + K$ and operator L and K are defined as:

$$L(\mathbf{X}) = -jk_0 \int_S \left(I + \frac{\nabla \nabla}{k_0^2} \right) G(\mathbf{r}, \mathbf{r}') \cdot \mathbf{X}(\mathbf{r}') dS', \quad (2)$$

$$K(\mathbf{X}) = P.V. \int_S \nabla G(\mathbf{r}, \mathbf{r}') \times \mathbf{X}(\mathbf{r}') dS', \quad (3)$$

where $G(\mathbf{r}, \mathbf{r}') = e^{-jk_0|\mathbf{r}-\mathbf{r}'|}/4\pi|\mathbf{r}-\mathbf{r}'|$ denotes the free-space Green's function with wavenumber k_0 and P.V. stands for the Cauchy principal value integration.

On the other hand, IBC can give the following relation between $\mathbf{J}$ and $\mathbf{M}$ as:

$$z_s z_0 \mathbf{J} = \hat{\mathbf{n}} \times \mathbf{M}. \quad (4)$$

$$\begin{bmatrix} C - \frac{1}{2} - \sqrt{z_s} & \hat{n} \times C(\sqrt{z_s}) - \frac{\sqrt{z_s}}{2} \hat{n} \times I + \hat{n} \times I \\ -\hat{n} \times C + \frac{1}{2} \hat{n} \times I - \sqrt{z_s} \hat{n} \times I & C(\sqrt{z_s}) - \frac{\sqrt{z_s}}{2} - 1 \end{bmatrix} \begin{bmatrix} \tilde{\mathbf{J}} \\ \tilde{\mathbf{M}} \end{bmatrix} = \begin{bmatrix} -\mathbf{e}^{inc} - \mathbf{j}^{inc} \\ (-\mathbf{e}^{inc} - \mathbf{j}^{inc}) \times \hat{n} \end{bmatrix}. \quad (5)$$

It should be noticed that z_s is the normalized surface impedance of the cavity. Combining equations (1) and (4) and replacing $\mathbf{J}$ and $\mathbf{M}$ with $\tilde{\mathbf{J}} = z_0 \mathbf{J}$ and $\tilde{\mathbf{M}} = \mathbf{M} / \sqrt{z_s}$ yields the C-SDIE equation (5).

According to the DG method, $\mathbf{J}$ and $\mathbf{M}$ are expanded with the RWG basis or half RWG basis $\mathbf{g}$. The electric or magnetic current continuity between nonconformal meshes is inherently guaranteed by translating the term in the integral equation involving operator L into the following four terms [14]:

$$\begin{aligned} \int_S \mathbf{g}_m \cdot L(\mathbf{g}_n) dS &= -jk_0 \int_S \int_{S'} \mathbf{g}_m \cdot \mathbf{g}_n G dS' dS \\ &+ \frac{j}{k_0} \int_S \nabla \cdot \mathbf{g}_m \int_{S'} \nabla' \cdot \mathbf{g}_n G dS' dS - \frac{j}{k_0} \int_S \nabla \cdot \mathbf{g}_m \int_{C'} \hat{\mathbf{t}}_n \\ &\cdot \mathbf{g}_n G dC' dS - \frac{j}{k_0} \int_C \hat{\mathbf{t}}_m \cdot \mathbf{g}_m \int_{S'} \nabla' \cdot \mathbf{g}_n G dS' dC, \end{aligned} \quad (6)$$

where m, n are the number of the edges, $\mathbf{g}$ denotes either RWG basis or half RWG basis, and $\hat{\mathbf{t}}$ denotes the unit normal vector of the boundary edges.

It is found that the BDP used in [10] does not work well for a partially coated cavity. We here apply a simple and efficient preconditioner DSP to DG-C-SDIE. The entries of DSP matrix are:

$$P_{ij} = \begin{cases} M_{ij} & |r_i - r_j| \leq 0.3\lambda_0 \\ 0 & \text{elsewhere} \end{cases}, \quad (7)$$

where r_i and r_j represent the location of the testing and bases elements and λ_0 is the free space wavelength.

III. NUMERICAL RESULTS

In the following numerical experiments, the multi-level fast multipole algorithm (MLFMA) [15] is applied to reduce CPU time and memory requirement.

A. Coating scheme

To efficiently coat the cavities, we first should select a suitable normalized surface impedance z_s of the coating. According to [4], when the EM wave is vertically incident on the coating, the surface impedance z_s of the coating can be equivalent as:

$$z_s = \sqrt{\frac{\epsilon_r}{\mu_r}} \tanh(jk_0 d \sqrt{\epsilon_r \mu_r}), \quad (8)$$

where d is the thickness of the coating, ϵ_r and μ_r are relative permittivity and relative permeability respectively, and k_0 is the wave number in free space. The impedance boundary condition typically assumes a small tangential electric field and, therefore, the equivalent impedance at vertical incidence is generally regarded as a boundary approximation. The effectiveness of the impedance boundary condition has been extensively studied [14, 16]. To make a preliminary assessment of the wave-absorbing performance of coatings with different thicknesses, the following reflection coefficient formula for a

normally incident semi-infinite space is commonly used for estimation:

$$R = \left| \frac{1 - z_s}{1 + z_s} \right|. \quad (9)$$

In this paper, a real EM absorbing material with relative permittivity of (18.4, -1.09j) and relative permeability of (1.6, -1.67j) is used. The relation between reflectivity R and thickness of the coat d is shown in Fig. 1, where λ is the wavelength of absorbing material. The thickness of 0.117λ , 0.17λ and 0.21λ are chosen respectively as three coating cases (I, II, III) of the absorbing capacity from weak to strong. The normalized surface impedances are equivalent as (0.2819, 0.175j), (0.4936, 0.1302j) and (0.5927, -0.0396j), respectively.

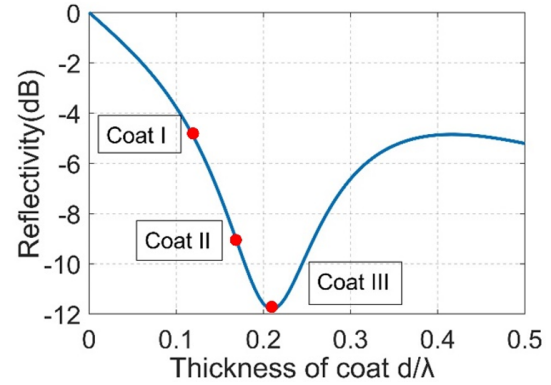


Fig. 1. Relationship between reflectivity and the thickness of a coating with relative permittivity of (18.4, -1.09j) and relative permeability of (1.6, -1.67j), where λ is the wavelength in absorbing material.

Our coating scheme is as follows. When the frequency of the EM wave is higher than the cutoff frequency of the cavity dominant mode, the EM wave can enter the cavity. We coat the cavity interior surface area directly illuminated by the EM wave incident in all concerned angles. To further reduce the coated region, we can neglect the part which is illuminated by the EM wave incident in a small angle range and coat the part which is illuminated by the EM wave incident from a large incidence angle range.

When the frequency of the EM wave is lower than the cutoff frequency of the cavity, the EM wave cannot enter the cavity. Therefore, we coat the external surface of the cavity.

B. Performance of DSP preconditioner

The numerical example is a partially coated rectangular cavity whose dimension is $1m \times 1m \times 2m$ and

Table 1: Comparison of numerical performance of NoP, BDP, DSP for the cavity in Fig. 2, where the degree of freedom is 5709

	Memory for PreC (MB)	CPU Time for PreC (s)	Iteration Number	Iteration Time (s)	Total Time (s)
NoP	—	—	76	23.14	23.14
BDP	3927	1468.89	21	10.85	1479.74
DSP	277	6.62	16	5.79	12.41

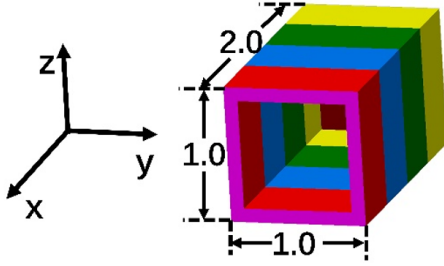


Fig. 2. Rectangle cavity with coated interior surface (impedance $z_s = 0.2819 + 0.175j$) and perfect electric conducting (PEC) external surface.

thickness is 0.1 m, shown in Fig. 2. The interior surface is an IBC boundary with impedance $z_s = 0.2819 + 0.175j$ and the external surface is a perfect electric conducting (PEC) material. The cavity is illuminated by a plane wave of 300 MHz, propagating in the x -direction with the electric field polarized in the z -direction. The cavity is divided into 11 subdomains to speed up the inverse of BDP preconditioner. Each subdomain is discretized independently with mesh size $\lambda/10$. The degree of freedom (DoF) of the cavity is 5709, with 3366 DoFs for the external cavity surface and 2343 DoFs for the interior cavity surface. The numerical performance of no preconditioner (NoP), BDP and DSP are shown in Table 1. The bistatic RCS and iteration convergence history are shown

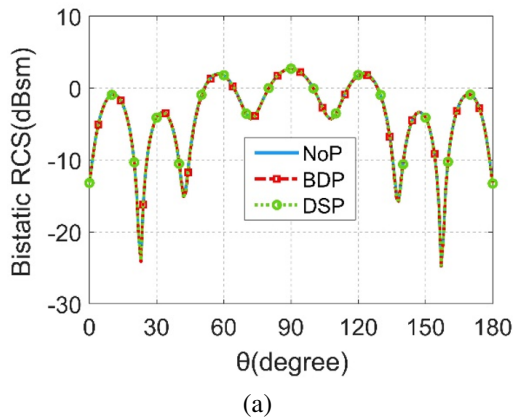


Fig. 3. Continued.

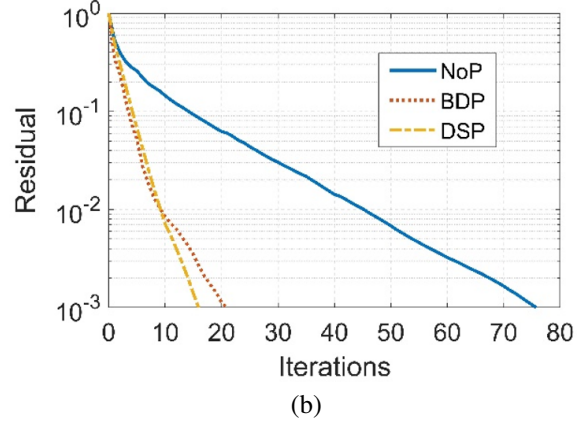


Fig. 3. (a) Bistatic VV-polarized RCS and (b) iteration convergence history for the cavity in Fig. 2, applying different preconditioners.

in Fig. 3. From Table 1 and Fig. 3, we can observe that, compared with BDP, DSP has a little faster convergence speed but saves considerable time and memory for constructing the preconditioner matrix. Obviously, DSP is more efficient when dealing with the coated cavity problem with larger DoFs.

C. Coated straight cavity

The EM scattering properties of a partially coated straight cavity are first investigated. Let us consider a rectangular cavity with an interior surface of $1m \times 4m \times 8m$ and a thickness of 0.1 m, as shown in Fig. 4. The surface impedance of coating is chosen as $(0.2819, 0.175j)$. The RCS calculation angle is $60 \sim 90$ degrees in the θ -direction and $\phi = 0$.

For a straight cavity, the entire cavity interior surface can be illuminated directly by the EM wave from our incident angle range. According to the surface area illuminated by EM waves in different incident angular

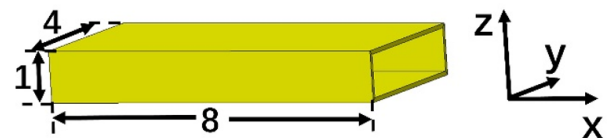


Fig. 4. Parameters of the rectangular cavity (unit: meter).

ranges, the interior surface of the cavity can be divided into four surface regions, as shown in Fig. 5 (a). The blue region is illuminated by an incident angle range of 60~90 degrees, the azure region by an incident angle range of 70~90 degrees, the green area by an incident angle range of 80~90 degrees and the yellow area by a much smaller incident angle range. For the straight cavity, coating the directly illuminated area is equivalent to coating the entire interior surface. According to our coating approach, three schemes are presented to further reduce the coated area, as shown in Fig. 5 (b). Scheme I coats the three regions of the blue, azure and green area; Scheme II coats the two regions of the blue and azure area; Scheme III only coats the blue region. We calculate the monostatic VV-polarization RCS of the PEC cavity, the cavity in which the total interior surface is coated (total coating), and the cavity coated by Scheme I, II and III at frequency $f = 600$ MHz.

Let us take the total coating as an example. The DoF of the cavity is 245709 with 128406 DoFs for the external cavity surface and 117303 DoFs for the interior cavity surface. The results are shown in Fig. 6. From Fig. 6, we see Scheme I has similar RCS reduction with the total coating. Scheme II also has comparable RCS reduction with the total coating, except for a little higher RCS at 80~90 degrees. Scheme III has similar RCS reduction with total coating at the angle range of 60~75 degrees,

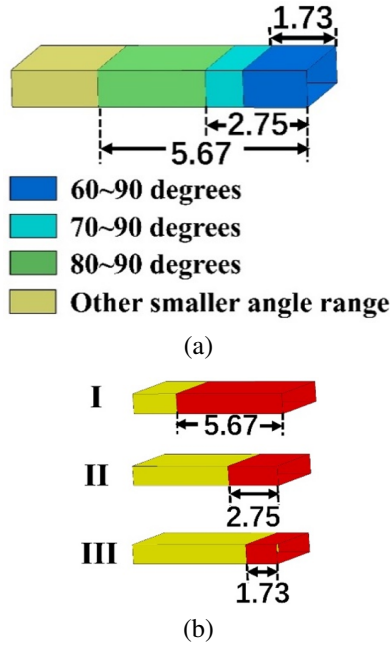


Fig. 5. The four regions illuminated by the EM wave from different angular ranges in the θ direction and (b) the three partial coating schemes (red area is coated and the unit is meter).

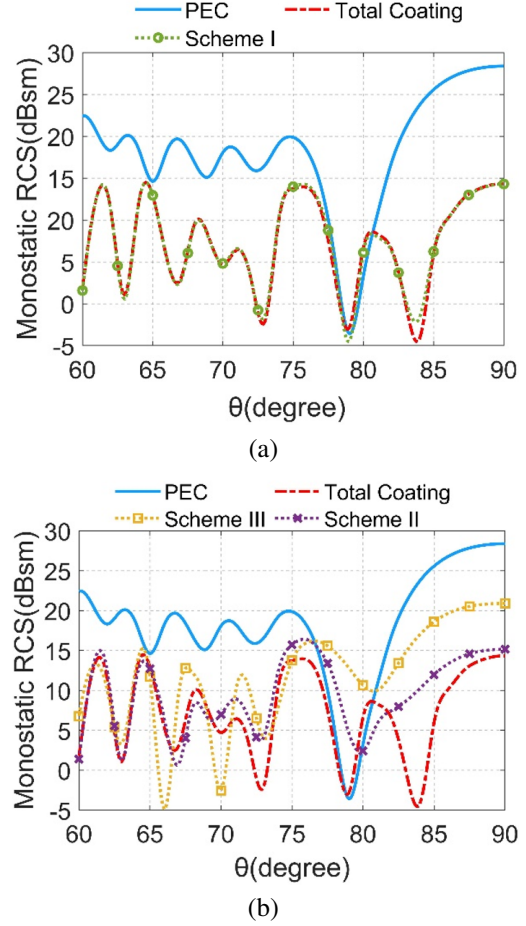


Fig. 6. (a) Monostatic VV-polarization RCS of PEC cavity in which the total interior surface is coated and (b) cavity coated by three schemes in Fig. 5 (b), at a frequency of 600 MHz.

but less RCS reduction at the angle range of 75~90 degrees, as we expected.

Next, we calculate the scattering from the above coated rectangular cavity at frequency of 15 MHz lower than the cavity dominant mode cutoff frequency of 37.5 MHz. The monostatic RCS of this coated cavity are shown in Fig. 7. As we expect, coating the interior surface of cavity cannot reduce RCS at all. For this case, coating the external surface of the cavity is the only way to reduce RCS.

D. Coated S-shaped cavity

We further investigate the EM scattering properties of the coated S-shaped cavity, which is widely used in the air inlet of stealth aircraft. As shown in Fig. 8, the model is a double S-shaped cavity with thickness 0.1 m and the specific parameters of the interior surface are as follows. The two S-shaped regions are 4.5 and 3.5 m long and offset distances in bending direction are 1.4 and 0.6

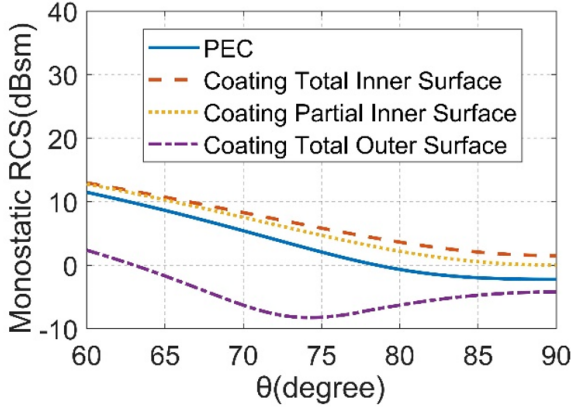


Fig. 7. Monostatic VV-polarization RCS of a PEC cavity in which the total inner and partial inner surface is coated, and a cavity in which the total outer surface is coated, at frequency of 15 MHz.

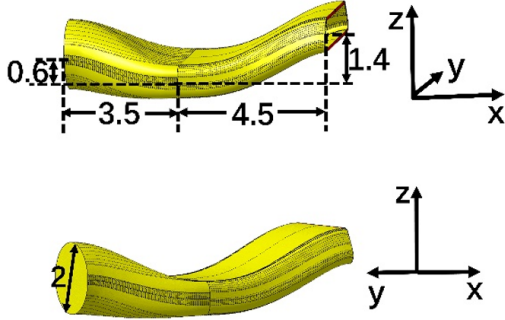


Fig. 8. Parameters of double-S shaped cavity (unit: meter).

m long; The bottom is a round surface with radius of 2 m while the opening is trapezoid with a median of 3.75 m and height of 0.94 m. The surface impedance of the coat is $(0.2819, 0.175j)$. The RCS calculation angle is $60 \sim 120$ degrees in the θ direction and $\varphi = 0$.

Based on the analysis of a coated rectangular cavity, the interior surface of the S-shaped cavity can be divided into different regions as shown in Fig. 9 (a). The yellow region is not illuminated directly by the EM wave; the dark green and light green regions are illuminated by incident angles of $60 \sim 82$ degrees; the red, blue, pink and purple regions are illuminated by an incident angle of $82 \sim 90$, $82 \sim 100$, $82 \sim 110$ and $82 \sim 120$ degrees; the orange region by a much smaller incident angle range. According to our coating approach, we have the following coating schemes. Scheme I only coats the region illuminated directly by the EM wave. Since the asymmetric structure of the S-shaped cavity makes the EM wave incident from large angles ($82 \sim 120$ degrees) illuminate a small region on the upper interior surface and the EM

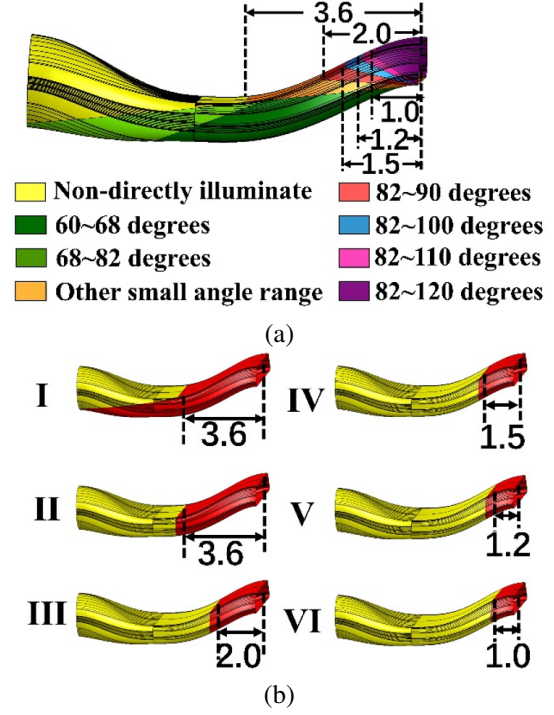


Fig. 9. (a) Eight regions illuminated by an EM wave from different angular ranges in the θ direction and (b) six partial coating schemes (red area is coated and the unit is meter).

wave incident from small angles ($60 \sim 82$ degrees) illuminates a large region on the lower interior surface, Scheme II neglects the region on the lower surface and only coats the region corresponding to $82 \sim 120$ degrees. Based on Scheme II, we further neglect the orange, red, blue and pink area in turn, corresponding to Schemes III, IV, V and VI, respectively. These schemes are shown in Fig. 9 (b). The monostatic VV-polarization RCS from the PEC cavity, the total coating and the cavity coated by the above-mentioned schemes at 600 MHz are calculated. Taking the total coating as an example, the DoF of the cavity is 200893, with 104653 DoFs for the external cavity surface, and 96240 DoFs for the interior cavity surface. We here only show the monostatic VV-polarization.

RCS of Scheme I and III in Fig. 10 and use the RCS from the PEC cavity and the total coating cavity as a comparison. It can be seen that Scheme I has nearly the same RCS reduction as total coating. Scheme III also has comparable RCS reduction as total coating, except for a little higher RCS at $65 \sim 82$ degrees, as we expected. To better demonstrate the effect of the coated area saving for reducing RCS, we defined the average reduction of RCS as:

$$\overline{RCS}_{reduce} = \left(\sum_{i=1}^{N_{obs}} (RCS_{PEC}^i - RCS_{IBC}^i) \right) / N_{obs}, \quad (10)$$

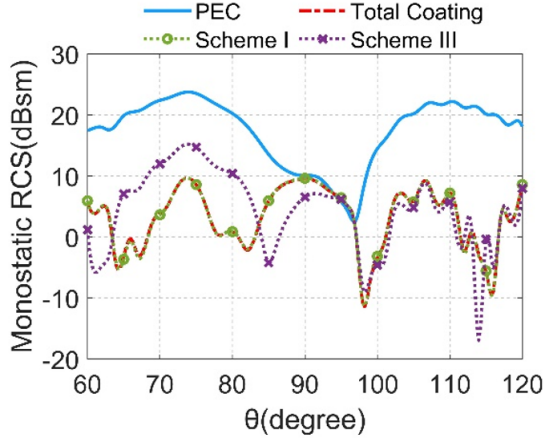


Fig. 10. Monostatic VV-polarization RCS of PEC cavity in which the total interior surface is coated, and cavity coated by Scheme I and III in Fig. 9 (b), at a frequency of 600 MHz.

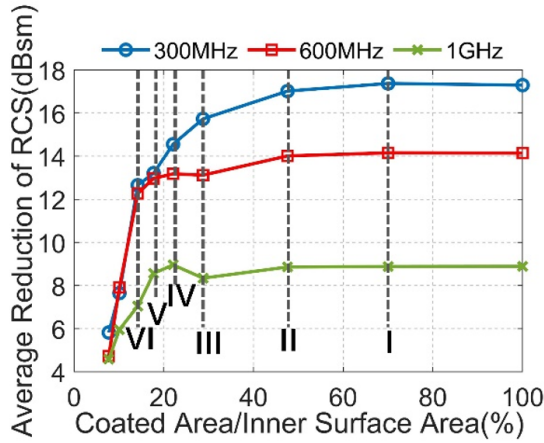


Fig. 11. Relationship between average reduction of the monostatic VV-polarization RCS and coated area ratio. Frequency is altered from 300 MHz to 1 GHz.

where N_{obs} denotes the number of observation angles and 121 is used in this example, RCS_{PEC}^i is the RCS of PEC cavity and RCS_{IBC}^i is the RCS of the coated cavity. The relationship between the coated area ratio and the average reduction of RCS is shown as a red line in Fig. 11 and the points corresponding to Scheme I~VI is marked. Clearly, coating the directly illuminated area (Scheme I) guarantees maximum (reliability for) RCS decrease and saves about 30% coating area while neglecting the area illuminated by a few angles (Scheme II~VI) saves about 50~85% area respectively and maintains enough RCS reduction.

E. Feasibility for different frequency

In this part, the feasibility of our partial coating scheme for different frequencies is investigated. We

employ the same impedance (0.2819, 0.175j), and set the different frequencies as 300 MHz, 600 MHz and 1 GHz; other parameters are the same as those in section IIID. The relationship between average reduction of the monostatic VV-polarization RCS and the coated area ratio for different coatings are shown in Fig. 11. As we see, the coated region directly illuminated (Scheme I) in the whole incident angle range has almost the same RCS reduction as the fully coated cavity for all frequencies. Moreover, we see Scheme II and III, which do not coat the interior surface illuminated by a small incident angle range, can also reduce enough RCS. All these demonstrate the applicability of our coating scheme.

F. Complex and large cavity with partial coating

Finally, a complex and large cavity with partial coating is presented to show the performance of our partial coating scheme. We consider a real inlet of a stealth aircraft, a three S-shaped cavity, whose opening is a parallelogram of 1157*966 mm and bottom is a circle with radius of 0.5 m, as shown in Fig. 12. This cavity is 97λ in length. The surface impedance of the coating is (0.4936, 0.1302j). We calculate the monostatic RCS in the incident angle range of 60~120 degrees in the θ -direction and 0 degrees in the ϕ -direction at a frequency of 4 GHz.

According to our partial coating scheme, a specific coating scheme is as follows. Coating the area of cavity directly illuminated in an incident angle range from 60 to 120 degrees, designated by the red area shown in Fig. 13. This coating scheme can save 11.1% coated area compared to total coating. The monostatic VV-polarization RCS of the PEC cavity, the total coating cavity and the

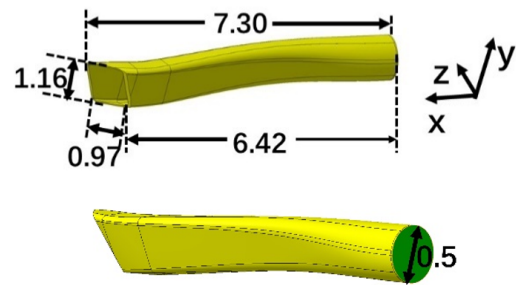


Fig. 12. Parameters of a realistic air inlet (unit: meter).

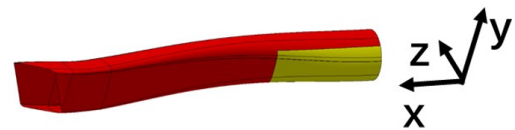


Fig. 13. Partial coating scheme for cavity in Fig. 12 (red area is coated).

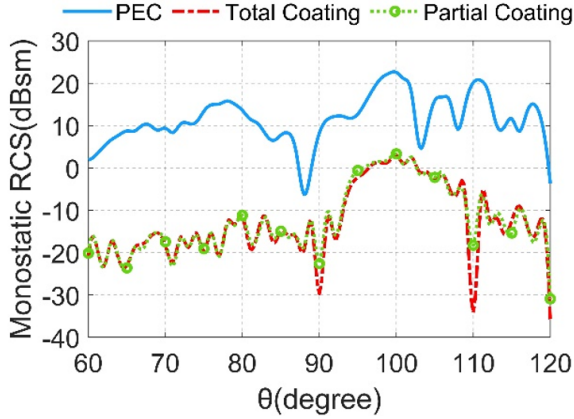


Fig. 14. Monostatic VV-polarization RCS of PEC cavity in which the total interior surface is coated, and cavity partially coated by the scheme in Fig. 13, at a frequency of 4 GHz.

cavity coated by our scheme are calculated. Take the total coating as an example. DoF of the cavity is 2828576 with 1464133 DoFs for the external cavity surface, and 1364443 DoFs for the interior cavity surface. RCS are shown in Fig. 14, and we can clearly see that our coating scheme can significantly reduce RCS and achieve almost the same results as the total coating.

IV. CONCLUSION

Scattering from partially coated cavities is studied efficiently and accurately by the improved DG-C-SDIE. The following useful conclusions can be obtained from the numerical experiments:

1. Coating the cavity interior surface region directly illuminated by the EM wave incident in whole angles can efficiently reduce backscattering and obtain almost the same RCS reduction as the total coating cavity.
2. The cavity interior surface illuminated by the EM wave incident in a small angle range can be further neglected and is not coated to further reduce the coated area with a little loss of RCS reduction.
3. For a working frequency lower than the cut-off frequency of the cavity, coating the interior cavity surface cannot reduce RCS. Instead, coating the external cavity surface is the only way to reduce the backscattering of the cavity.

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