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# Single Stage Power Control Strategy for Nanogrid Connected Solar PV System

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## Abstract

The focus on reduction of converter size in nanogrid connected Photo-Voltaic (PV) systems increases to obtain the various advantages. This paper proposes a single stage nanogrid connected solar PV system that works with DC-AC inverter along with the closed loop control strategy to send the maximum power from the PV array. The proposed scheme controls the power delivery of the PV system to nanogrid in a single stage power conversion along with the incorporated Maximum Power Point Tracking (MPPT) controller. The decoupled control strategy has been designed and developed for the proposed system using synchronous reference frame theory. The developed hardware prototype in the laboratory has been used to validate MATLAB/Simulink simulation results. The simulation and experimental results have been observed

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for dynamic changes in irradiation by maintaining the Total Harmonic Distortion (THD) within the standard limits while operating at unity power factor. The successful validation of the proposed control strategy shows its usefulness in real-time nanogrid applications.

**Keywords:** Single stage power conversion, nanogrid, decoupled control strategy, maximum power point tracking, and DC-AC inverter.

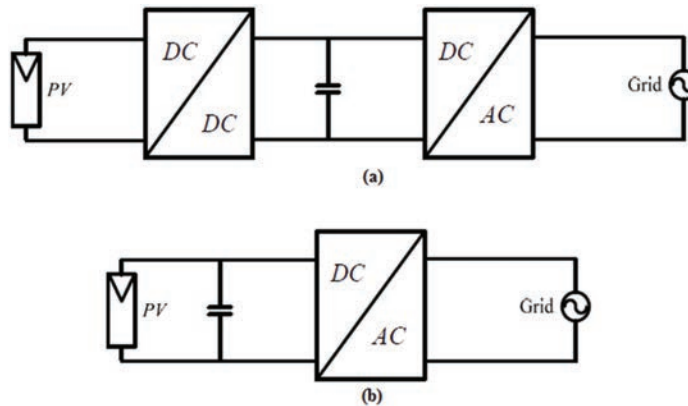
## 1 Introduction

The increased demand of electrical energy, increased prices of fossil fuels such as petroleum, coal, natural gas, heavy crude oil etc. makes us to use Renewable Energy Sources (RES). Among the available RES, solar energy has the benefits of inexhaustible, clean and environment friendly [1]. Now a day, the advanced power electronic controllers are used to harness maximum power output and which has less operational and maintenance costs. As many of the electrical loads are operated with AC power supply, AC transmission is preferred over DC transmission. Therefore, DC power produced in solar PV systems are converted into suitable AC power supply and fed to utility grid. To regulate the terminal voltage, constant frequency and also for feeding electrical power to utility grid it requires suitable power electronic controllers [2].

A grid-connected solar PV system has become most popular application of solar energy [3]. MPPT technique is applied in grid-connected photovoltaic system to enhance the power fed to grid [4]. Solar PV system, which uses a multi-level inverter for feeding power is little complex in developing its control algorithm [5]. Modelling of Grid-connected PV system control strategy with three phase three level inverter is examined [6]. The capacitor voltage balance limit is a major constraint in a multilevel converter based energy storage system as discussed in [7]. Hence, multi-level inverter with space-vector PWM method is implemented [8]. The variation of middle point voltage control for three level VSI is implemented with modified PWM control scheme [9]. Further, the three level voltage source inverters are implemented with randomized voltage vector switching scheme [10]. The single stage three phase solar PV system has improved maximum power tracking capability [11]. The single stage grid-connected solar PV systems have benefits such as simple structure, high efficacy, etc. Now a day, grid-connected solar PV system becomes more popular in most of the applications [12, 13]. However, the control objectives such as the synchronization with the utility

voltage, peak power point tracking and reduction of harmonics in grid current need to be considered [14]. Various peak power point tracking techniques are implemented for uniform insolation and non-uniform insolation on PV array to extract global peak power from PV modules [15]. The most popular and widely used MPPT method for PV modules is Perturb & Observe (P&O), because of its less complexity. By using conventional series-parallel connection, various topologies are proposed to link local PV generation system to grid [16].

Now a day, single stage grid-connected solar PV systems becomes more popular as it does not require any storage systems such as batteries and also the capital cost is reduced [17]. The extraction of maximum power from PV modules becomes essential because of its high initial cost and less span of life. There are mainly two types of power electronic configurations in grid-connected PV systems such as single stage and two stage power conversions between PV array and grid. In two stage configuration of a grid connected solar PV system, DC-DC converter succeeded by DC-AC inverter [18, 19] as given in Figure 1(a). The DC-DC converter is used to get the variable resistance across the PV array terminals and also to give the required DC voltage at input of DC-AC inverter stage. The power transferred to the grid is achieved by generating the sinusoidal voltages and currents from the inverter. It has lower reliability, larger filter size and less efficiency. The harmonics reduction is achieved by implementing a modified Sinusoidal Pulse Width Modulation (SPWM) based switching strategy [20]. Further, nano-grids with the renewable energy sources are complex systems and also depend on



**Figure 1** Grid-connected converter configuration (a) Conventional two-stage nanogrid configuration (b) Single-stage nanogrid configuration.

different dynamic sources to feed various applications. The International Renewable Energy Agency (IRENA) report highlights about nano-grid such as improving the resilience and reliability in power system by integrating the wind and solar power generation sources. The renewable nano-grids are different from the main grid, because the islands and remote areas can also get the reliable electricity from renewable nano-grids [21].

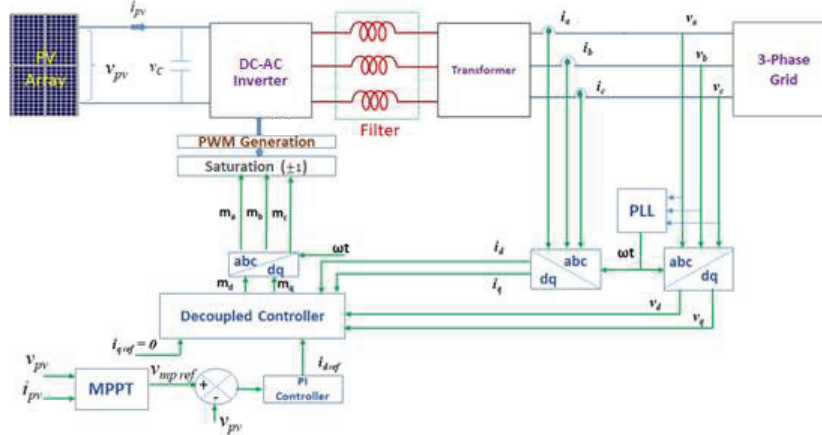
This paper elaborates the single stage grid connected solar PV system and its control technique. The single stage power converters block diagram is as shown in Figure 1(b). The objectives of two stage power converters need to be fulfilled by single stage power converter. Single stage conversion can be achieved by two methods with low cost and higher efficiency. One of the methods is cascaded inverter with step-up transformer, and the other is PV array with sufficiently large voltage. The large voltage at PV array suffers due to partial shading of the PV modules. Among the two methods, inverter followed by the transformer configuration is most commonly used for the single stage power conversion. DC-AC voltage source inverters have been widely used in industry for single stage DC to AC power conversion.

After detailed literature survey, the problems identified are as follows: (i) two stage power conversion, (ii) less power delivery between PV array and grid, (iii) dedicated power converter for MPPT. Hence, the objectives of the proposed method to address afore mentioned problems are as follows: (i) Developing the single stage power converter, (ii) Design of controller for maximum power delivery between solar PV array and grid, (iii) Incorporating MPPT in single stage converter.

This paper organized as follows: The proposed single stage grid connected solar PV system is discussed in Section 2. The results and discussions are elaborated in Section 3. Finally, Section 4 is given with the conclusions remarks.

## **2 Proposed Single Stage Grid Connected Solar PV System**

The proposed nanogrid connected PV system is given in Figure 2. It consists of the solar PV array at its input terminal, MPPT controller to obtain maximum power, DC-AC inverter to integrate solar PV array with the nanogrid, transformer matches the inverter output voltage with voltage of the nanogrid, a small size inductive filter to minimize nanogrid current harmonics, and DC link capacitor to reduce DC link voltage ripples. The decoupled current control strategy is used to control the power flow between the PV array and nanogrid.



**Figure 2** Block diagram for proposed nanogrid configuration.

Further, DC-AC Inverter is connected to PV array through suitable DC link capacitor. The output of PV array voltage is maintained constant by using MPPT controller. The closed loop control strategy for DC-AC Inverter is used to feed the power to the nanogrid. It is known that the obtained power from the solar PV array varies widely and it is directly fed to the nanogrid. The flowchart for the proposed control strategy is given in Figure 3.

The voltage controller is used to generate the reference for the control of active power. It compares the voltage at DC bus and the generated reference voltage by MPPT controller by following the Proportional Integral (PI) controller. The PI controller’s output provides the active current component ( $i_{dref}$ ) and reactive power control achieved with the reactive current component. Depending on the reactive power requirement to the nanogrid, the component of reactive current is adjusted. If nanogrid is not demanding for reactive power then the component of reactive current is adjusted to zero.

The inverter output voltage is the vector sum of voltages at AC grid and the series drop of across the filter. The voltage equation for three phases is given below:

$$v_a = V_{ga} + Ri_a + L \frac{di_a}{dt} \tag{1}$$

$$v_b = V_{gb} + Ri_b + L \frac{di_b}{dt} \tag{2}$$

$$v_c = V_{gc} + Ri_c + L \frac{di_c}{dt} \tag{3}$$

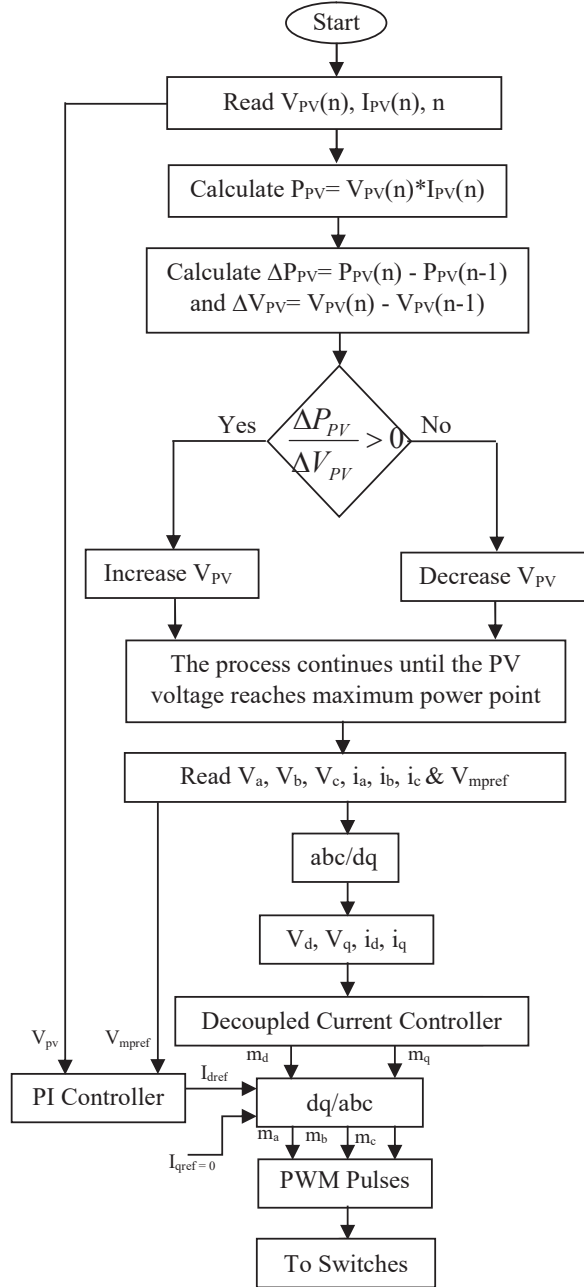


Figure 3 Flowchart for the proposed control strategy.

where,  $V_{ga}$ ,  $V_{gb}$ , and  $V_{gc}$  are the nanogrid voltages.  $V_a$ ,  $V_b$ , and  $V_c$  are the output voltages at inverter. The total inductance between the nanogrid and inverter is  $L$ .  $R$  is the resistance of filter.

The above equations are altered into  $dq$  quantities by using synchronously rotating  $dq$  reference frame. The modified equations are;

$$v_d = V_{gd} + Ri_d + L \frac{di_d}{dt} - \omega Li_q \quad (4)$$

$$v_q = V_{gq} + Ri_q + L \frac{di_q}{dt} - \omega Li_d \quad (5)$$

Where,  $V_{gd}$  and  $V_{gq}$  are the AC system voltage components,  $i_d$  and  $i_q$  are the AC system current components and  $v_d$  and  $v_q$  are inverter output voltages in the  $dq$ -frame.

The relationship between  $v_d$  and  $m$  is given as:

$$v_d = m_d \frac{v_{dc}}{2} \quad (6)$$

$$v_q = m_q \frac{v_{dc}}{2} \quad (7)$$

The true and reactive power delivered to the AC side is,

$$p = \frac{3}{2}(V_{gd}i_d + V_{gq}i_q) \quad (8)$$

$$q = \frac{3}{2}(-V_{gd}i_q + V_{gq}i_d) \quad (9)$$

Under steady state and nanogrid synchronized condition,  $V_{gq} = 0$ . The above equations can be written as,

$$p = \frac{3}{2}V_{gd}i_d \quad (10)$$

$$q = \frac{-3}{2}V_{gd}i_q \quad (11)$$

Therefore, based on Equations (10) and (11),  $p$  and  $q$  are controller by  $i_d$  and  $i_q$  respectively.

$$i_{dref} = \frac{2}{3V_{gd}}p_{ref} \quad (12)$$

$$i_{qref} = \frac{2}{3V_{gd}}q_{ref} \quad (13)$$

From the above equations, if  $i_d \approx i_{dref}$  and  $i_q \approx i_{qref}$ , then the injection of desired reactive power and true power into the nanogrid. The fast tracking of reference is achieved using the voltage and current controllers. From these, it can say that  $p$  and  $q$  are independently controlled by their respective reference commands.

In Equations (4) and (5),  $i_d$  and  $i_q$  are state variables,  $v_d$  and  $v_q$  are control inputs, and  $V_{gd}$  and  $V_{gq}$  are disturbance inputs. Due to the presence of  $L\omega_0$  terms in Equations (4) and (5), the dynamics of  $i_d$  and  $i_q$  are coupled. To decouple, dynamics of  $m_d$  and  $m_q$  are determined as follows

$$m_d = \frac{2}{v_{dc}}(u_d + V_{gd} - \omega L i_q) \quad (14)$$

$$m_q = \frac{2}{v_{dc}}(u_q + V_{gq} - \omega L i_d) \quad (15)$$

Where  $u_d$  and  $u_q$  are two new control inputs. Whereas,

$$L \frac{di_d}{dt} = u_d - R i_d \quad (16)$$

$$L \frac{di_q}{dt} = u_q - R i_q \quad (17)$$

The above equations describe two decoupled first-order linear systems. Based on Equations (16) and (17),  $i_d$  and  $i_q$  can be controlled by  $u_d$  and  $u_q$  respectively. Figure 4 shows a block diagrammatic representation of the  $d - q$  axis current controllers of the solar PV system connected to nanogrid in which  $u_d$  and  $u_q$  are the outputs of two corresponding compensators. The error  $e_d = i_d^{ref} - i_d$  processed from d-axis compensator and provides  $u_d$ . Then, based on Equation (14),  $u_d$  contributes to  $m_d$ . Similarly, the error  $e_q = i_q^{ref} - i_q$  processed from q-axis compensator and provides  $u_q$ . Then based on Equation (15),  $u_q$  contributes to  $m_q$ . The voltage source inverter amplifies  $m_d$  and  $m_q$  by a factor of  $\frac{v_{dc}}{2}$  and generates  $V_d$  and  $V_q$  that in turn control  $i_d$  and  $i_q$ . As all signals are DC quantities, a simple PI compensator is sufficient to enable tracking of DC reference commands. These two PI regulators compensate any existing error by generating appropriate control voltages  $u_d$  and  $u_q$ , which are altered back to phase quantities ( $m_a, m_b, m_c$ ) and are given as inputs to PWM block to generate the command signals to the switches of inverter.



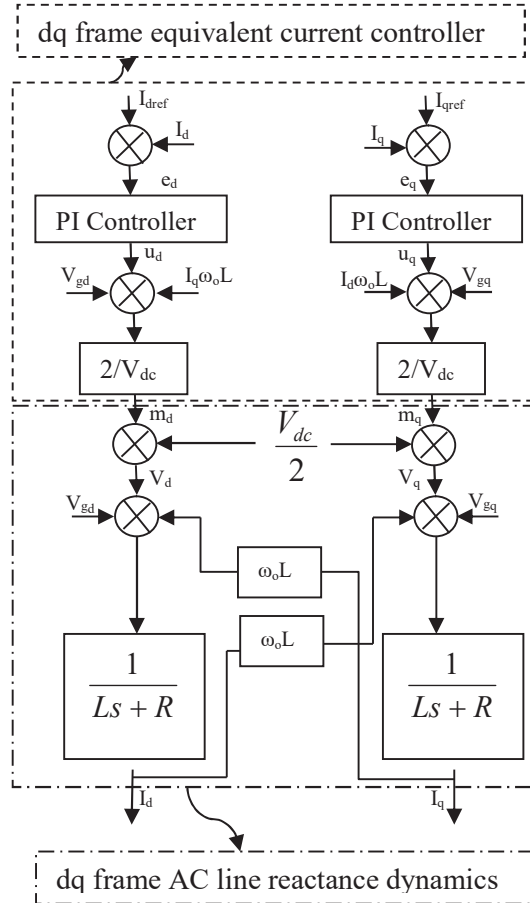


Figure 4 Control block diagram of a current controlled voltage source inverter.

### 3 Results and Discussion

The MATLAB/Simulink simulated and hardware results of the single stage nanogrid connected solar PV system with proposed controller using dSPACE 1103 controller board are presented. The proposed system performance obtained through simulation are validated with the experimentally developed 500 W prototype consist of three PV modules connected in series with each module rating of 166.6 W. The PV array is designed based on the module parameter values presented in Table 1. The complete experimental set up is shown Figure 5. It consists of DC-AC inverter along with the gate drive circuit

**Table 1** Specifications of PV module

Parameter	Value
PV module power, $P_o$	166.8 W
PV module open circuit voltage, $V_{oc}$	46 V
PV module short circuit current, $I_{sc}$	4.7 A
Maximum module voltage, $V_{mp}$	37.5 V
Maximum module current, $I_{mp}$	4.45 A
Number of modules in series	3

**Table 2** Specifications of DC-AC inverter

Parameter	Value
Inverter DC voltage	0–600 V DC
Output AC voltage	415 V AC
Output current	30 A max
Switching frequency	20 kHz max
Ambient temperature	45°C
Cooling method	Forced air cooled

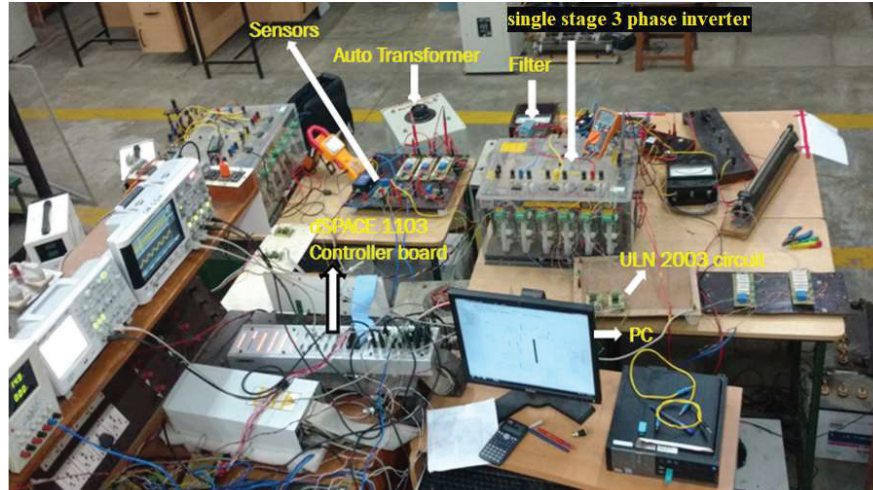
**Table 3** Specifications of LEM voltage sensor LV 25-P

Parameter	Value
Primary involved potential, $V_{PN}$	DC/AC 500 V
Supply voltage, $V_c$	12 V to 15 V DC
Primary nominal RMS current, $I_{PN}$	0 to 10 mA
Conversion ratio, $K_N$	2500:1000
Secondary nominal RMS current, $I_{SN}$	25 mA

**Table 4** Specifications of LEM current sensor LA 55-P

Parameter	Value
Primary nominal RMS current, $I_{PN}$	50 A DC/AC
Conversion ratio, $K_N$	1:1000
Secondary nominal RMS current, $I_{SN}$	50 mA

with the specifications as given in Table 2 along with the autotransformer to set the nanogrid voltage to 62V and L-filter. For sensing the currents and voltages, LEM based voltage and current sensors are used with the specifications as given in Tables 3 and 4. The proposed controller performance is observed by implementing in dSPACE 1103 real time controller software desk. For modelling and code generation MATLAB/Simulink Real-Time Workshop by the Math Works and the Real-Time Interface by dSPACE are installed. Control desk software from dSPACE is used as a front end and serves to



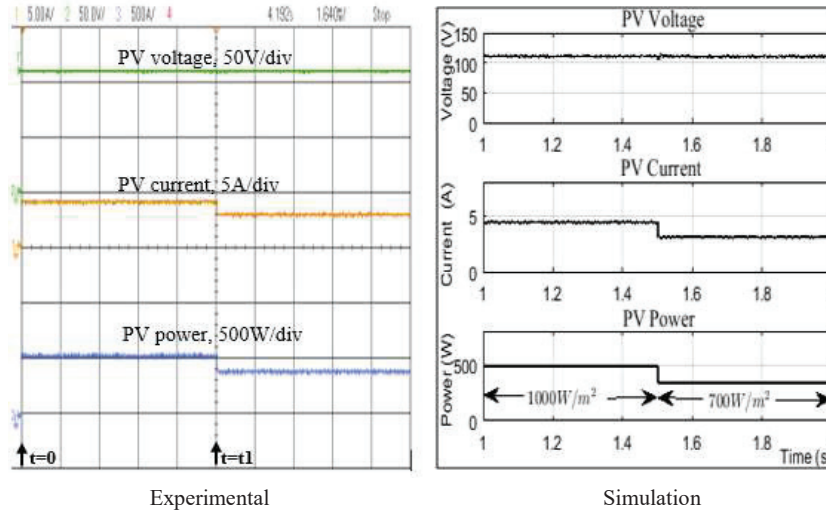
**Figure 5** Experimental set-up of single stage nanogrid connected solar PV system.



**Figure 6** dSPACE 1103 controlled board.

provide instrumentation, parameterization, measurements, and experiment control as shown in Figure 6. The three phase nanogrid parameters are line to line voltage of 62.46 V and frequency of 50 Hz. The simulation results obtained for  $1000 \text{ W/m}^2$  is from 1 to 1.5 s and  $700 \text{ W/m}^2$  is from 1.5 to 2 s. The experimental results are obtained for  $1000 \text{ W/m}^2$  is from  $t = 0$  to  $t_1$  sec and  $700 \text{ W/m}^2$  is from  $t > t_1$  sec. The temperature is kept constant at  $25^\circ\text{C}$ .

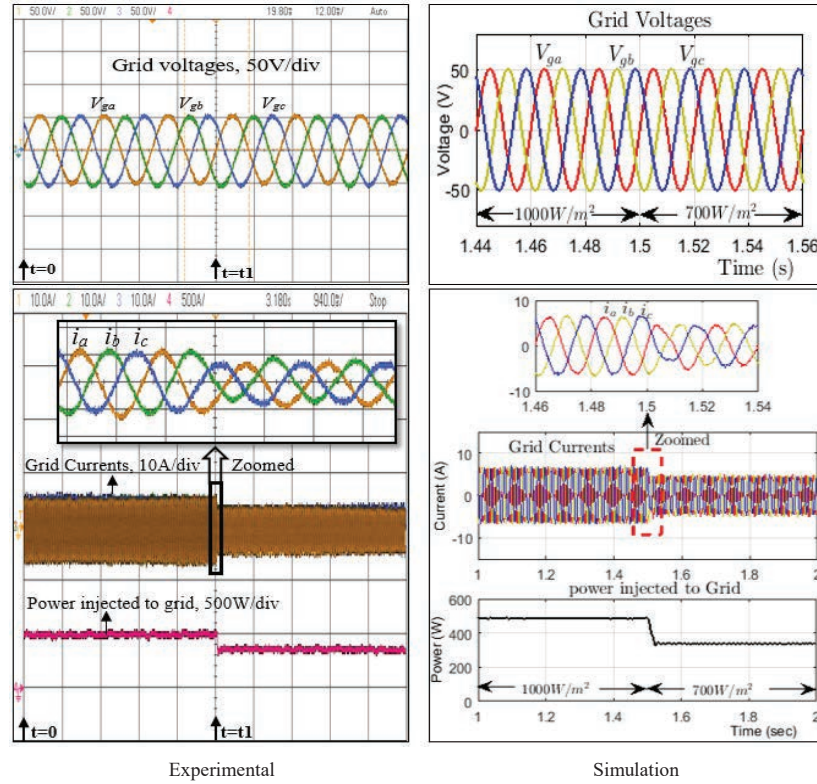
PV systems equipment such as step-up transformers and electrical cables are not sources of electromagnetic interference because of their low-frequency (50 Hz) of operation and PV panels themselves do not emit EMI.



**Figure 7** Dynamic response of PV parameters for step change in irradiance from  $1000 \text{ W/m}^2$  to  $700 \text{ W/m}^2$ .

The only component of a PV array that may be capable of emitting EMI is the inverter. Hence, synthesizing a stepped output voltage allows reduction in harmonic content of voltage waveform. This results in increasing the quality of output waveform and reducing the size and cost of the output filter. The staircase output voltage can improve the quality of the output voltage and reduce the voltage stress ( $dv/dt$ ) on switching components. This can remedy the problem associated with Electro Magnetic Interference (EMI) problems.

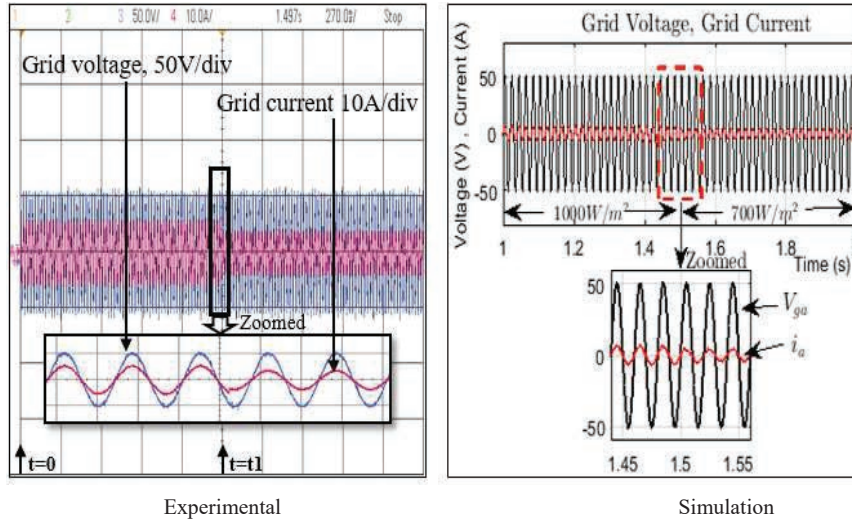
The dynamic response of PV array parameters for step change in irradiance from  $1000 \text{ W/m}^2$  to  $700 \text{ W/m}^2$  is shown in Figure 7. A solar PV module with a maximum power of  $166.6 \text{ W}$ , maximum power voltage of  $37.5 \text{ V}$ , and maximum power current of  $4.7 \text{ A}$  is emulated in the  $2 \text{ kW}$  Chroma solar simulator 62020H-150S-Chroma-2014. Three such modules are connected in series and feed to the developed grid connected PV system. The required irradiances are adjusted as considered in the simulations results ( $700 \text{ W/m}^2$  and  $1000 \text{ W/m}^2$ ) by using the solar simulator knobs. For irradiance of  $1000 \text{ W/m}^2$ : PV Voltage, PV Current and PV Power respectively are  $112.5 \text{ V}$ ,  $4.45 \text{ A}$  and  $500 \text{ W}$  for simulation and experimentation. For irradiance of  $700 \text{ W/m}^2$ , the PV array voltage, PV current and PV power respectively are  $111.8 \text{ V}$ ,  $3.2 \text{ A}$  and  $358 \text{ W}$  for simulation and  $111 \text{ V}$ ,  $3.2 \text{ A}$  and  $355 \text{ W}$  for experimentation as shown in Figure 7. The grid phase voltage, current and power injected to nanogrid are shown in Figure 8. For irradiance



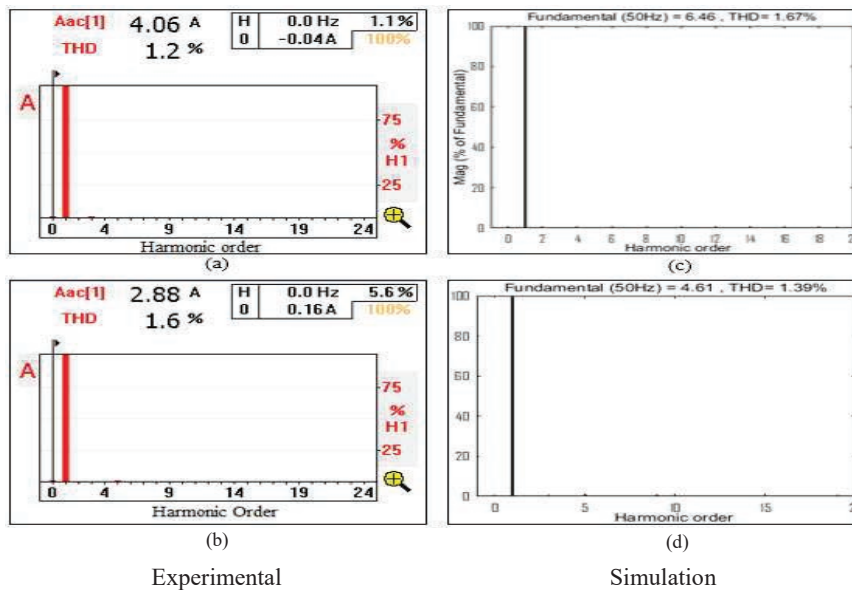
**Figure 8** Dynamic responses of grid side values for change in irradiance from  $1000 \text{ W/m}^2$  to  $700 \text{ W/m}^2$ .

of  $1000 \text{ W/m}^2$ : grid phase voltage, current and power injected to nanogrid respectively are  $36.06 \text{ V}$ ,  $4.57 \text{ A}$  and  $495 \text{ W}$  for simulation and  $36.1 \text{ V}$ ,  $4.06 \text{ A}$  and  $440 \text{ W}$  for experimentation. For irradiance of  $700 \text{ W/m}^2$ : grid phase voltage (rms), current (rms) and power injected to nanogrid respectively are  $36.06 \text{ V}$ ,  $3.26 \text{ A}$  and  $353 \text{ W}$  for simulation and  $36.1 \text{ V}$ ,  $2.88 \text{ A}$  and  $312 \text{ W}$  for experimentation as shown in Figure 8.

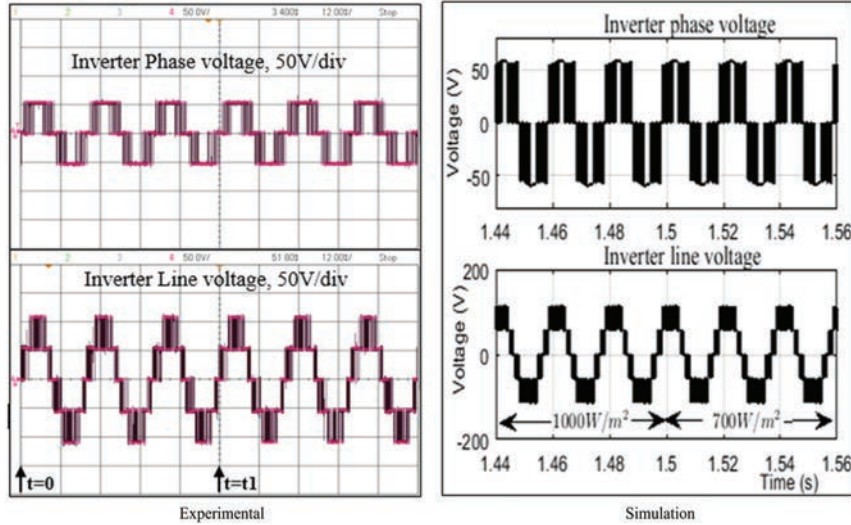
Figure 9 shows the phase current and phase voltage in nanogrid. The phase current and phase voltage is almost in phase, which authenticates that the system is operating at unity power factor. Figure 10(a) shows the experimental result of the nanogrid current THD at  $1000 \text{ W/m}^2$  and Figure 10(b) for THD at  $700 \text{ W/m}^2$  respectively and the corresponding current THD value are  $1.2\%$  and  $1.6\%$  respectively. The simulation results for the nanogrid current THD at  $1000 \text{ W/m}^2$  and  $700 \text{ W/m}^2$  are given in Figure 10(c) and 10(d)



**Figure 9** Voltage and current waveforms of the nanogrid at unity power factor.



**Figure 10** Nanogrid current THD for step change in irradiance from 1000 W/m<sup>2</sup> to 700 W/m<sup>2</sup>.



**Figure 11** DC-AC inverter output voltages for step change in irradiance from 1000 W/m<sup>2</sup> to 700 W/m<sup>2</sup>.

**Table 5** Simulation and Experimental efficiency calculations with power loss analysis

Simulation	Experimentation
<ul style="list-style-type: none"> <li>• <b>From t = 1 to 1.5 Sec (1000 W/m<sup>2</sup>):</b> DC: V = 112.5 V, I = 4.45 A, P = 500 W AC: V<sub>ph</sub> = 36.06 V, I = 4.57 A, P = 495 W Efficiency = 495/500 = 99%</li> <li>• <b>Suddenly injecting the grid current:</b> At t = 1.5 Sec (700 W/m<sup>2</sup>) DC: V = 111.8 V, I = 3.2 A, P = 358 W AC: V<sub>ph</sub> = 36.06 V, I = 3.26 A, P = 353 W Efficiency = 353/358 = 98.6%</li> </ul>	<ul style="list-style-type: none"> <li>• <b>From t = 0 to t<sub>1</sub> Sec (1000 W/m<sup>2</sup>):</b> DC: V = 112.5 V, I = 4.45 A, P = 500 W AC: V<sub>ph</sub> = 36.1 V, I = 4.06 A, P = 440 W Efficiency = 440/500 = 88%</li> <li>• <b>Suddenly injecting the grid current:</b> At t = t<sub>1</sub> Sec (700 W/m<sup>2</sup>) DC: V = 111.8 V, I = 3.2 A, P = 355 W AC: V<sub>ph</sub> = 36.1 V, I = 2.88 A, P = 312 W Efficiency = 312/355 = 87.8%</li> </ul>

\*In experimentation, the losses are due to the switching losses and I<sup>2</sup>R drop.

respectively. The corresponding current THD values are 1.67% and 1.39% respectively. From above results, it is found that the values of current THD in nanogrid is well within the international standard (i.e., <5%). Figure 11 shows the DC-AC inverter output voltages for step change in irradiance from 1000 W/m<sup>2</sup> to 700 W/m<sup>2</sup>. The simulation and experimental efficiency calculations with power loss analysis is given in Table 5. Further, the proposed control logic has been compared with the existed and highlighted the advantages as given in Table 6.

**Table 6** Performance parameters comparison of proposed configuration with existed configurations

Parameters	Configurations		
	[12]	[13]	Proposed
Power level	Medium Power	HighPower	HighPower
Power Factor	0.98	Nearly Unity	Unity
Efficiency	NA	95%	99%
Inductor	1	4	3
Capacitor	1	3	1
Switch	4	4	6
Symmetrical/Unsymmetrical	Unsymmetrical	Unsymmetrical	Symmetrical
Frequency	20 KHz	4 KHz	5 KHz
Implementation of MPPT	P&O	Fuzzy	P&O
PV Module	BP-340	NA	Chroma solar simulator 62020H-150S
Grid Current THD	6.05%	<5%	1.2%

## 4 Conclusion

This paper presented the control scheme for the single stage power conversion for the nanogrid applications. The experimental results had been validated with simulation results for power delivery between solar PV array and nanogrid under different solar irradiance. The MPPT controller had been successfully incorporated in single stage converter with DC-AC inverter. Hence, the overall size of the converter was reduced. Decoupled control strategy was designed and developed for the proposed system using synchronous reference frame theory. The decoupled control strategy provides independent control over injection of true power to the nanogrid with the simultaneous tracking of the peak power from the PV array. The performance of the proposed scheme had been assessed through MATLAB/Simulink simulations and verified experimentally by means of dSPACE 1103 controller in laboratory. The successful validation of results shows the ability of the proposed control strategy to feed the peak PV power to the nanogrid. The values of current THD in nanogrid maintains very well within the range of IEEE-519 recommended standards while operating the system at unity power factor.

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